



EE-602
STATISTICAL SIGNAL PROCESSING

PRESENTATION ON
CONSTRAINED LEAST SQUARES
&
NON-LINEAR LEAST SQUARES

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CONSTRAINED LEAST SQUARES

At times we come across the **LS** problems where unknown parameters must be constrained. Assuming parameter θ is subject to $r < p$ independent constraints, then we summarize the constraints as

$$\mathbf{A}\theta = \mathbf{b}$$

Where \mathbf{A} is a $r \times p$ matrix and \mathbf{b} is a known $r \times 1$ vector.

For instance, let $p=2$ and one parameter is negative of other, then the constraint would be

$$\theta_1 + \theta_2 = 0$$

Then we have

$$\mathbf{A} = \begin{bmatrix} 1 & 1 \end{bmatrix} \quad \text{and} \quad \mathbf{b} = 0$$

The condition for the constraints to be independent is that the matrix \mathbf{A} should be full rank (equal to r). There are only $p-r$ independent parameters.

Contd...

To find the **Least Squares Estimator (LSE)** subject to linear constraints, we use **Lagrangian multiplier**.

We determine θ_c (c denotes the constrained LSE) by minimizing the Lagrangian

$$J_c = (\mathbf{x} - \mathbf{H}\theta)^T (\mathbf{x} - \mathbf{H}\theta) + \lambda^T (\mathbf{A}\theta - \mathbf{b})$$

Where λ is a $r \times 1$ vector of Lagrangian multipliers.
Expanding above expression, we have

$$J_c = \mathbf{x}^T \mathbf{x} - 2\theta^T \mathbf{H}^T \mathbf{x} + \theta^T \mathbf{H}^T \mathbf{H} \theta + \lambda^T \mathbf{A} \theta - \lambda^T \mathbf{b}$$

taking the gradient w.r.t. θ , we have

$$\frac{\partial J_c}{\partial \theta} = -2\mathbf{H}^T \mathbf{x} + 2\mathbf{H}^T \mathbf{H} \theta + \mathbf{A}^T \lambda$$

Contd...

Setting it to zero, we have

$$\begin{aligned}\hat{\theta}_c &= (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x} - \frac{1}{2} (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T \lambda \\ &= \hat{\theta} - (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T \lambda \quad (\text{i})\end{aligned}$$

Where $\hat{\theta}$ is the unconstrained LSE.

Now, to find λ , we impose the constrain

$$\mathbf{A} \theta = \mathbf{b}$$

So that

$$\mathbf{A} \theta_c = \mathbf{A} \hat{\theta} - \frac{1}{2} \mathbf{A} (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T \lambda = \mathbf{b}$$

Hence

$$\frac{1}{2} \lambda = [\mathbf{A} (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T]^{-1} (\mathbf{A} \hat{\theta} - \mathbf{b})$$

Substituting in equation (i) gives

$$\hat{\theta}_c = \hat{\theta} - (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T [\mathbf{A} (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T]^{-1} (\mathbf{A} \hat{\theta} - \mathbf{b}) \quad (\text{ii})$$

Where

$$\hat{\theta} = (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x}$$

Example : Constrained Signal

If the signal model is

$$\mathbf{s}[\mathbf{n}] = \begin{cases} \theta_1 \\ \theta_2 \\ 0 \end{cases}$$

and we observe $\{\mathbf{x}[0], \mathbf{x}[1], \mathbf{x}[2]\}$, then the observation matrix is

$$\mathbf{H} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{bmatrix}$$

The signal vector is

$$\mathbf{s} = \mathbf{H}\boldsymbol{\theta} = \begin{bmatrix} \theta_1 \\ \theta_2 \\ 0 \end{bmatrix}$$

Example contd...

The unconstrained **LSE** is

$$\hat{\theta} = (H^T H)^{-1} H^T \mathbf{x} = \begin{bmatrix} \mathbf{x}[0] \\ \mathbf{x}[1] \end{bmatrix}$$

and the signal estimate is

$$\hat{\mathbf{s}} = \mathbf{H} \hat{\theta} = \begin{bmatrix} \mathbf{x}[0] \\ \mathbf{x}[1] \\ 0 \end{bmatrix}$$

Now assume that we know a priori that

$$\theta_1 = \theta_2$$

thus we have

$$[1 \quad -1] \theta = 0$$

so that

$$\mathbf{A} = [1 \quad -1] \text{ and } \mathbf{b} = 0$$

Example contd...

Now noting that

$$\mathbf{H}^T \mathbf{H} = \mathbf{I}$$

so from equation (ii) we have

$$\begin{aligned}\hat{\boldsymbol{\theta}}_c &= \hat{\boldsymbol{\theta}} - \mathbf{A}^T [\mathbf{A} \mathbf{A}^T]^{-1} \mathbf{A} \hat{\boldsymbol{\theta}} \\ &= [\mathbf{I} - \mathbf{A}^T [\mathbf{A} \mathbf{A}^T]^{-1} \mathbf{A}] \hat{\boldsymbol{\theta}}\end{aligned}$$

on solving the above expression, we have

$$\hat{\boldsymbol{\theta}}_c = \begin{bmatrix} \frac{1}{2} (x[0] + x[1]) \\ \frac{1}{2} (x[0] + x[1]) \end{bmatrix}$$

and the constrained signal becomes

$$\hat{\mathbf{s}}_c = \mathbf{H} \hat{\boldsymbol{\theta}}_c = \begin{bmatrix} \frac{1}{2} (x[0] + x[1]) \\ \frac{1}{2} (x[0] + x[1]) \\ 0 \end{bmatrix}$$

Nonlinear Least Squares

Least Square Procedure estimates model parameters θ by minimizing the Least Square Error Criterion:

$$J = (\mathbf{x} - s(\theta))^T (\mathbf{x} - s(\theta))$$

$s(\theta)$ = Signal model for \mathbf{x} , with its dependence on θ

Methods that can reduce complexity of Problem for determining nonlinear LSEs :

- 1. Transformation of parameters*
- 2. Separability of parameters*

Nonlinear Least Squares

1. Transformation of parameters

*one-to-one transformation of θ that produces a **linear** signal model in the new space*

$$\alpha = \mathbf{g}(\theta)$$

\mathbf{g} : p -dimensional function of θ whose inverse exists. If a \mathbf{g} can be found so that

$$s(\theta) = s(\mathbf{g}^{-1}(\alpha)) = \mathbf{H}\alpha$$

then the signal model will be linear in α .

Nonlinear Least Squares

Thus the **nonlinear** LSE of θ by

$$\theta' = g^{-1}(\alpha')$$

Where : $\alpha' = (H^T H)^{-1} H^T x$

Approach: *minimization can be carried out in any transformed space that is obtained by a one-to-one mapping & then converted back to original space.*

Nonlinear Least Squares

2. Separability of parameters

A separable signal model has the form

$$s = H(\alpha)\beta$$

where

$$\theta = \begin{bmatrix} \alpha \\ \beta \end{bmatrix} = \begin{bmatrix} (p - q) \times 1 \\ q \times 1 \end{bmatrix}$$

$H(\alpha)$ is an $N \times q$ matrix dependent on α .

This model is linear in β but nonlinear in α . The LS Error may be minimized with respect to β and thus reduced to a function of α only.

Nonlinear Least Squares

Since

$$J(\alpha, \beta) = (\mathbf{x} - H(\alpha)\beta)^T (\mathbf{x} - H(\alpha)\beta)$$

the β that minimizes J for a given α is

$$\beta' = (H^T(\alpha)H(\alpha))^{-1} H^T(\alpha)\mathbf{x}$$

And the resulting LS error is

$$J(\alpha, \beta') = \mathbf{x}^T [I - H(\alpha)(H^T(\alpha)H(\alpha))^{-1} H(\alpha)^T] \mathbf{x}$$

The Problem now reduces to a **maximization** of

$$\mathbf{x}^T H(\alpha)(H^T(\alpha)H(\alpha))^{-1} H(\alpha)^T \mathbf{x}$$

over α .

Nonlinear Least Squares

Example : Sinusoidal Parameter Estimation

For a sinusoidal signal model

$$s[n] = A \cos(2\pi f_o n + \phi)$$

*It is desired to estimate the amplitude A , where $A > 0$, phase ϕ .
The frequency f_o is assumed known.*

LSE is obtained by minimizing

$$J = \sum_{n=0}^{N-1} (x[n] - A \cos(2\pi f_o n + \phi))^2$$

Nonlinear Least Squares

Over A & ϕ , a nonlinear LS Problem. However, because

$$A \cos(2\pi f_o n + \phi) = A \cos(\phi) \cos(2\pi f_o n) - A \sin(\phi) \sin(2\pi f_o n)$$

If we let

$$\alpha_1 = A \cos(\phi)$$
$$\alpha_2 = -A \sin(\phi)$$

Then the signal model becomes

$$s[n] = \alpha_1 \cos(2\pi f_o n) + \alpha_2 \sin(2\pi f_o n)$$

In the Matrix form this is $s = H\alpha$

Nonlinear Least Squares

where

$$H = \begin{bmatrix} 1 & 0 \\ \cos(2\pi f_o n) & \sin(2\pi f_o n) \\ \vdots & \vdots \\ \cos(2\pi f_o (N - 1)) & \sin(2\pi f_o (N - 1)) \end{bmatrix}$$

Which is now linear in the new parameters.

The LSE of α is

$$\alpha' = (H^T H)^{-1} H^T x$$

and to find θ' we must find the inverse transformation $g^{-1}(\alpha)$.

Nonlinear Least Squares

This is

$$A = \sqrt{(\alpha_1^2 + \alpha_2^2)}$$

$$\phi = \tan^{-1} \left(\frac{-\alpha_2}{\alpha_1} \right)$$

so that the nonlinear LSE for this problem is given by

$$\theta' = \begin{bmatrix} A' \\ \phi' \end{bmatrix}$$

$$\theta' = \begin{bmatrix} \sqrt{(\alpha_1^2 + \alpha_2^2)} \\ \tan^{-1} \left(\frac{-\alpha_2}{\alpha_1} \right) \end{bmatrix}$$

where $\alpha' = (H^T H)^{-1} H^T x$