

**EE-602**

**STATISTICAL SIGNAL PROCESSING**

TERM PAPER REPORT ON

**CONSTRAINED AND NON-LINEAR LEAST SQUARES**



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## CONSTRAINED LEAST SQUARES

At times we come across the LS problems where unknown parameters must be constrained. Assuming parameter  $\theta$  is subject to  $r < p$  independent constraints, then we summarize the constraints as

$$A\theta = b$$

Where  $A$  is a  $r \times p$  matrix and  $b$  is a known  $r \times 1$  vector.

For instance, let  $p=2$  and one parameter is negative of other, then the constraint would be

$$\theta_1 + \theta_2 = 0$$

Then we have

$$A = [1 \ 1]$$

$$\text{and } b = 0$$

The condition for the constraints to be independent is that the matrix  $A$  should be full rank (equal to  $r$ ). There are only  $p-r$  independent parameters.

To find the Least Squares Estimator (LSE) subject to linear constraints, we use Lagrangian multiplier.

We determine  $\theta_c$  ( $c$  denotes the constrained LSE) by minimizing the Lagrangian

$$J_c = (x - H\theta)^T(x - H\theta) + \lambda^T(A\theta - b)$$

Where  $\lambda$  is a  $r \times 1$  vector of Lagrangian multipliers.

Expanding above expression, we have

$$J_c = x^T x - 2\theta^T H^T x + \theta^T H^T H \theta + \lambda^T A \theta - \lambda^T b$$

Taking the gradient w.r.t.  $\theta$ , we have

$$\frac{dJ}{d\theta} c = -2H^T x + 2H^T H \theta + A^T \lambda$$

Setting it to zero, we have

$$\begin{aligned} \hat{\theta}_c &= (H^T H)^{-1} H^T x - \frac{1}{2} (H^T H)^{-1} A^T \lambda \\ &= \hat{\theta} - \frac{1}{2} (H^T H)^{-1} A^T \lambda \end{aligned} \quad (i)$$

Where  $\hat{\theta}$  is the unconstrained LSE.

Now, to find  $\lambda$ , we impose the constrain

$$A\theta = b$$

So that

$$A\theta_c = A\hat{\theta} - \frac{1}{2} A(H^T H)^{-1} A^T \lambda = b$$

Hence

$$\frac{1}{2} \lambda = [A(H^T H)^{-1} A^T]^{-1} (A\hat{\theta} - b)$$

Substituting in equation (i) gives

$$\hat{\theta}_c = \hat{\theta} - (H^T H)^{-1} A^T [A(H^T H)^{-1} A^T]^{-1} (A\hat{\theta} - b) \quad (ii)$$

Where

$$\hat{\theta} = (H^T H)^{-1} H^T x$$

## EXAMPLE OF CONSTRAINED LEAST SQUARES

If the signal model is

$$s[n] = \begin{cases} \theta_1 \\ \theta_2 \\ 0 \end{cases}$$

and we observe  $\{x[0], x[1], x[2]\}$ , then the observation matrix is

$$H = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{bmatrix}$$

The signal vector is

$$s = H\theta = \begin{bmatrix} \theta_1 \\ \theta_2 \\ 0 \end{bmatrix}$$

The unconstrained LSE is

$$\hat{\theta} = (H^T H)^{-1} H^T x = \begin{bmatrix} x[0] \\ x[1] \end{bmatrix}$$

and the signal estimate is

$$\hat{s} = H\hat{\theta} = \begin{bmatrix} x[0] \\ x[1] \\ 0 \end{bmatrix}$$

Now assume that we know a priori that

$$\theta_1 = \theta_2$$

thus we have

$$[1 \ -1] \theta = 0$$

so that

$$A = [1 \ -1]$$

$$\text{and } b = 0$$

Now noting that

$$H^T H = I$$

so from equation (ii) we have

$$\begin{aligned} \hat{\theta}_c &= \hat{\theta} - A^T [A A^T]^{-1} A \hat{\theta} \\ &= [I - A^T [A A^T]^{-1} A] \hat{\theta} \end{aligned}$$

on solving the above expression, we have

$$\hat{\theta}_c = \begin{bmatrix} \frac{1}{2}(x[0] + x[1]) \\ \frac{1}{2}(x[0] + x[1]) \end{bmatrix}$$

and the constrained signal becomes

$$\hat{s}_c = H \hat{\theta}_c = \begin{bmatrix} \frac{1}{2}(x[0] + x[1]) \\ \frac{1}{2}(x[0] + x[1]) \\ 0 \end{bmatrix}$$

## NONLINEAR LEAST SQUARES

Least Square Procedure that estimates model parameters  $\theta$  by minimizing the Least Square Error Criterion:

$$J = (\mathbf{x} - \mathbf{s}(\theta))^T (\mathbf{x} - \mathbf{s}(\theta))$$

$\mathbf{s}(\theta)$  = Signal model for  $\mathbf{x}$ , with its dependence on  $\theta$ .

In the Linear Least Square problem the signal takes on special form  $\mathbf{s}(\theta) = \mathbf{H}\theta$ , which leads to the simple linear Least Square Estimator. In general,  $\mathbf{s}(\theta)$  can't be expressed in this manner but is an N-dimensional nonlinear function of  $\theta$ . In such a case the minimization of J becomes much more difficult, if not impossible. This type of nonlinear Least Square problem is often termed a *nonlinear regression problem*.

**Methods that can reduce complexity of Problem for determining nonlinear LSEs:**

1. Transformation of parameters
2. Separability of parameters

### 1. Transformation of parameters

One-to-one transformation of  $\theta$  that produces a **linear** signal model in the new space

$$\alpha = \mathbf{g}(\theta)$$

$\mathbf{g}$ : p-dimensional function of  $\theta$  whose inverse exists. If a  $\mathbf{g}$  can be found so that

$$s(\theta) = s(g^{-1}(\alpha)) = H\alpha$$

then the signal model will be linear in  $\alpha$ .

Thus the **nonlinear** LSE of  $\theta$  by

$$\hat{\theta} = g^{-1}(\hat{\alpha})$$

where:  $\hat{\alpha} = (H^T H)^{-1} H^T x$

**Approach:** minimization can be carried out in any transformed space that is obtained by a one-to-one mapping & then converted back to original space.

## 2 . Separability of parameters

A separable signal model has the form

$$s = H(\alpha)\beta$$

$$\theta = \begin{bmatrix} \alpha \\ \beta \end{bmatrix} = \begin{bmatrix} (p - q) \times 1 \\ q \times 1 \end{bmatrix}$$

$H(\alpha)$  is an  $N \times q$  matrix dependent on  $\alpha$ .

This model is linear in  $\beta$  but nonlinear in  $\alpha$ . The LS Error may be minimized with respect to  $\beta$  and thus reduced to a function of  $\alpha$  only.

Since

$$J(\alpha, \beta) = (x - H(\alpha)\beta)^T (x - H(\alpha)\beta)$$

the  $\beta$  that minimizes  $J$  for a given  $\alpha$  is

$$\hat{\beta} = (H^T(\alpha)H(\alpha))^{-1} H^T(\alpha)x$$

And the resulting LS error is

$$J(\alpha, \hat{\beta}) = x^T [ I - H(\alpha)(H^T(\alpha)H(\alpha))^{-1} H(\alpha)^T ] x$$

The Problem now reduces to a **maximization** of

$$\mathbf{x}^T \mathbf{H}(\alpha) (\mathbf{H}^T(\alpha) \mathbf{H}(\alpha))^{-1} \mathbf{H}(\alpha)^T \mathbf{x}$$

over  $\alpha$ . If, for instance,  $q=p-1$ , so that  $\alpha$  is a scalar, then a grid can possibly be used. This should be contrasted with the original minimization of a  $p$ -dimensional function.

### **Example: Sinusoidal Parameter Estimation**

For a sinusoidal signal model

$$s[n] = A \cos(2\pi f_0 n + \phi)$$

It is desired to estimate the amplitude  $A$ , where  $A > 0$ , phase  $\phi$ .

The frequency  $f_0$  is assumed known.

LSE is obtained by minimizing

$$J = \sum_{n=0}^{N-1} (x[n] - A \cos(2\pi f_0 n + \phi))^2$$

Over  $A$  &  $\phi$ , a nonlinear LS problem. However, because

$$A \cos(2\pi f_0 n + \phi) = A \cos(\phi) \cos(2\pi f_0 n) - A \sin(\phi) \sin(2\pi f_0 n)$$

If we let  $\alpha_1 = A \cos(\phi)$

$$\alpha_2 = -A \sin(\phi)$$

Then the signal model becomes

$$s[n] = \alpha_1 \cos(2\pi f_0 n) + \alpha_2 \sin(2\pi f_0 n)$$

In the Matrix form this is  $\mathbf{s} = \mathbf{H}\alpha$



Where,

$$H = \begin{bmatrix} 1 & 0 \\ \cos(2\pi f_0 n) & \sin(2\pi f_0 n) \\ \vdots & \vdots \\ \cos(2\pi f_0 (N-1)) & \sin(2\pi f_0 (N-1)) \end{bmatrix}$$

which is now linear in the terms of new parameters .

The LSE of  $\alpha$  is

$$\hat{\alpha} = (H^T H)^{-1} H^T X$$

and to find  $\hat{\theta}$  we must find the inverse transformation  $g^{-1}(\alpha)$ .

This is

$$A = (\alpha_1^2 + \alpha_2^2)^{1/2}$$

$$\phi = \tan^{-1} \left( \frac{\alpha_2}{\alpha_1} \right)$$

So that the nonlinear LSE for this problem is given by

$$\hat{\theta} = \begin{bmatrix} \hat{A} \\ \hat{\phi} \end{bmatrix}$$

$$\hat{\theta} = \begin{bmatrix} \sqrt{\alpha_1^2 + \alpha_2^2} \\ \tan^{-1} \left( \frac{\alpha_2}{\alpha_1} \right) \end{bmatrix}$$

Where:  $\hat{\alpha} = (H^T H)^{-1} H^T X$